**PUBLISHED** 

MAY 25, 1943.

BY A. P. C.

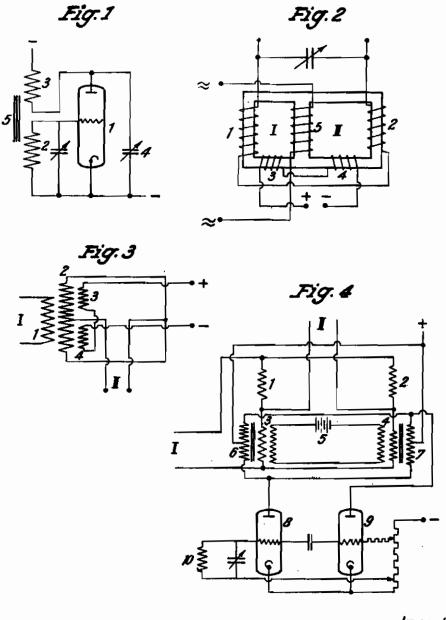
L L DE KRAMOLIN

RADIO APPARATUS

Filed March 11, 1940

Serial No. 323,450

7 Sheets-Sheet 1

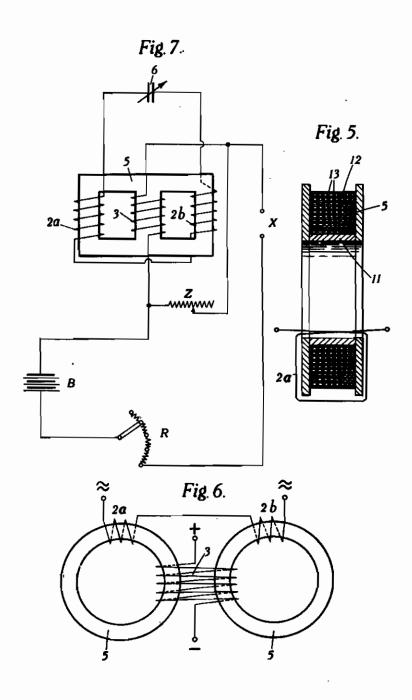


By stony, Eng. Hagen Attorneys PUBLISHED MAY 25, 1943. BY A. P. C.

L L DE KRAMOLIN
RADIO APPARATUS
Filed March 11, 1940

Serial No. 323,450

7 Sheets-Sheet 2



By your Enventor:

By your Enventor:

Attorneys

PUBLISHED

MAY 25, 1943.

BY A. P. C.

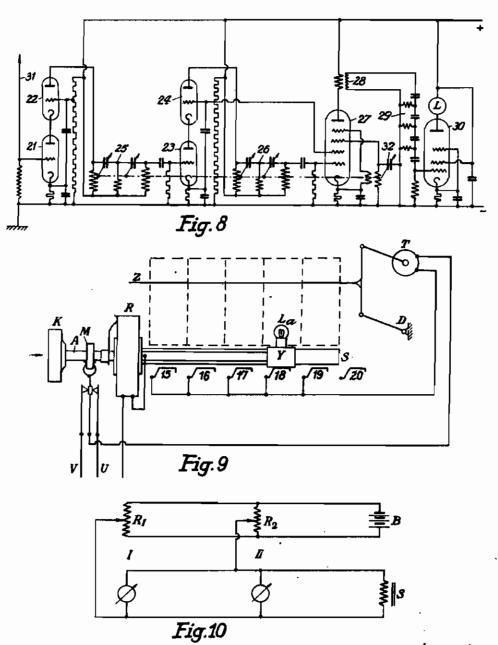
L L DE KRAMOLIN

RADIO APPARATUS

Filed March 11, 1940

Serial No. 323,450

7 Sheets-Sheet 3



Inventor: .

L.L. De Kramolin

By young, Emery Horpon

Attorneys

PUBLISHED .

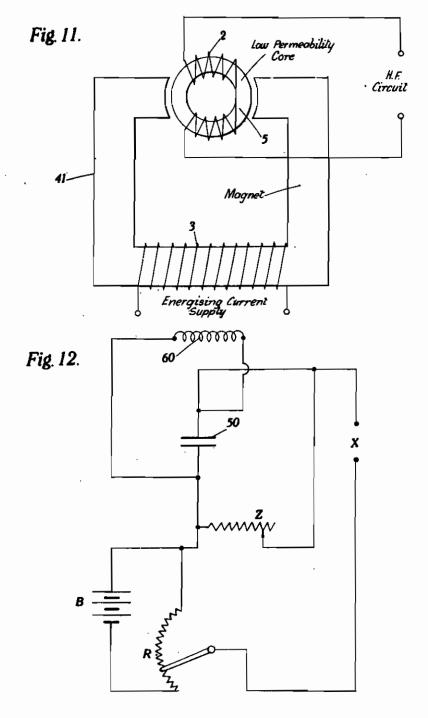
L L DE KRAMOLIN
RADIO APPARATUS

Serial No. 323,450

MAY 25, 1943. BY A. P. C.

Filed March 11, 1940

7 Sheets-Sheet 4



By four land the

PUBLISHED

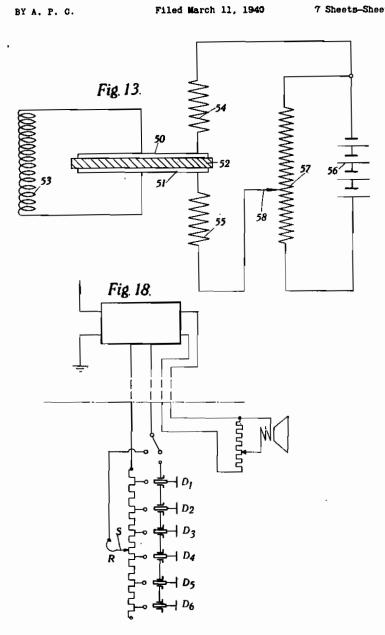
L. L. DE KRAMOLIN

Serial No. 323,450

MAY 25, 1943.

RADIO APPARATUS

7 Sheets-Sheet 5



PUBLISHED MAY 25, 1943.

BY A. P. C.

L L DE KRAMOLIN

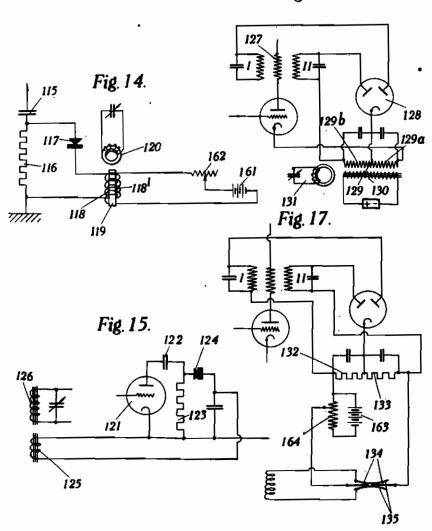
RADIO APPARATUS

Filed March 11, 1940

Serial No. 323,450

7 Sheets-Sheet 6

Fig. 16.



Inventor:
L.L. De Kramolin
By
Young, Eveny Ylangum
Attorneys

**PUBLISHED** 

MAY 25, 1943.

BY A. P. C.

L L DE KRÁMOLIN

RADIO APPARATUS

Filed March 11, 1940

Serial No. 323,450

7 Sheets-Sheet 7

Fig. 19.

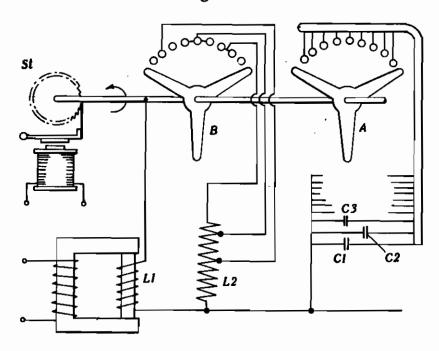
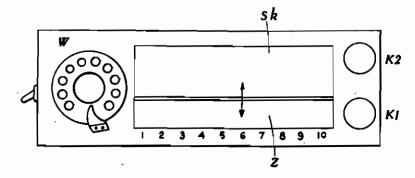


Fig. 20.



Inventor: LL. De Kramolin By Young, Energy Horna Attorneys

## ALIEN PROPERTY CUSTODIAN

## RADIO APPARATUS

Leon Ladislaus de Kramolin, Haarlem, Holland; vested in the Alien Property Custodian

Application filed March 11, 1940

This application is a continuation in part of the copending Application No. 749,088, filed October 18, 1934.

My invention relates particularly to a method of regulating radio receiving sets and the like 5 by means of inductance variations in iron cores.

As is well-known, with sufficiently fine subdivision of suitable types of iron alloys, it is possible to provide high-frequency coils with magnetic cores. If the cores are made displace- 10 able within the coils tuning can be effected by such displacement in a manner well-known per se. This variation of inductance, however, can also be obtained in a different manner, e. g. by varying the preliminary magnetisation of the 15core by means of a direct current. Such a method of varying the inductance offers a large number of possibilities of utilization, one very interesting instance of which comprises the provision of an automatic volume control.

In the simplest case, such an apparatus works as follows: An oscillatory circuit, tuned in some manner to the wave to be received is provided, the oscillations of which are applied to an audion. In the oscillatory circuit, a part or the 25 whole of the inductance is arranged on a coil of ferro-magnetic material e. g. Ferrocart. In the anode circuit of the audion, there is provided a simple filtering means, which separates the direct current component of the anode current 30 from the alternating current component. The alternating current component (modulation frequency) is supplied to the consumer, that is, a telephone or a further amplifier, and the direct current component is led through a second 35 winding on the ferromagnetic core of the tuning coil.

On increase of the audion current or decrease of the audion current (according as to whether grid or anode current rectification is employed), 40 the result can then be obtained that, if normal volume is employed, the coil is just tuned to the oscillation to be received. If the volume increases beyond an agreeable extent, then (assuming that working is being effected with the 45 detector tube at the lower bend) an increase of the anode current causes a decrease in the inductance of the coil, which detunes the receiving oscillatory circuit to such an extent that an agreeable volume is just restored. It is seen 50 that a simple volume control device can be assembled in this manner. In this case it is possible, not only to decrease the volume if the reception is too powerful, but also, quite simply if

turbance level would be too high in proportion, to obtain a decrease of volume. For this purpose, it is only necessary to balance the tuning accurately at a mean value of the volume. If the volume falls below this value, then by weakening the direct current component of the anode current of the detector tube, an increase of the inductance and thus detuning will occur, while with excessive volume or field strength of the receiver oscillation at the receiving station, the self-inductance becomes too small. In normal automatic volume control arrangements there often exists an undesirable relationship between the time constant of the regulating circuit and the regulating quality or other characteristics of the regulator. With the present construction and also in all other possibilities discussed here. however, it is possible regardless of the time constant, to suppress any other action between the regulating and regulated circuit by entirely decoupling the regulating circuit from the regulated circuit, e. g. by use well-known per se of a three-limb transformer or the like. If due to small asymmetries in the masses of the limbs or in the coil configuration, there should be no complete decoupling at the outset then a subsequent correction can be obtained either by displacement of individual convolutions or groups of convolutions or by readjusting the core by a slight subsequent deformation, say, by cutting off small particles or by deformation by pressure, which is quite possible in view of the relatively soft consistency of such cores.

Should the presence of a sharp resonance curve cause an excessive instability in the volume, any desired flattening of the regulating curve may be produced by increasing the damping of the circuit. It is also possible by the use of chain conductors to deform the resonance curve in such a manner that the desired regulating characteristic is produced. In order that the selectivity of the reception may not be impaired by such measures, individual invariable, more or less sharply tuned, circuits may be employed for the actual receiving selection and these circuits coupled to the detector tube by circuit variable in its tuning in accordance with the volume in the manner hereinbefore described. Of course, any desired number of high-frequency amplifying stages may also be connected in series with the detector tube, which stages may be influenced by the receiving field strength in a similar manner as has been herein described for a detector. The greater the number of stages in the volume is insufficient, in which case the dis- 55 which the influencing is effected, the greater of

course is the effect. In order, to avoid, particularly in the case of a number of circuits, displacing the resonance position of the whole receiving system within the frequency spectrum in accordance with the volume, it may be prefer- 5 able in two successive circuits or else within a single circuit, by reversal of the direction of current in the preliminary magnetisation windings on occurrence of the regulating action, to displace the resonance position of the part-circuits 10 of a system in the opposite direction in the frequency spectrum, so that the resultant resonance position remains unchanged. Since the regulating circuit is decoupled in relation to the regulated circuits, the same regulating current in 15 series-connection or parallel-connection may traverse the regulating windings of the transformers of a larger number of series-connected amplifier stages without any tendency of the successive amplifier stages to oscillate, arising. To 20 this end, exact decoupling is alone essential, and this can be ensured by any means well-known per se. In order also to prevent capacitive backcouplings through the regulating line, it is preferable to provide for good direct earthing or 25 earthing effected through adequate capacities. However, as is hereinafter shown, volume control is not only obtainable by means of detuning. A very simple alternative for instance, comprises automatically regulating the back-coupling, the 30 back-coupling being effected through one of the aforementioned ferromagnetic mass cores or the like.

Other regulating actions can also be obtained in accordance with the invention: Thus, for instance, it is well-known that with only weak reproduction an apparent displacement of the deep sounds occurs. Since, variation of the preliminary magnetisation causes variation of the inductance it is possible with low-frequency transformers or low-frequency sound monitoring arrangements, by means of filters etc., to ensure that the constants of these constructional elements automatically vary in such a manner that the percentage of deep sounds is increased or reduced as desired in relation to the percentage of high sounds, so that with changes of volume, effected for instance, by a manual regulator, provided in addition to the automatic regulator, the inductance of the low-frequency transformers or the low-frequency chokes or any kinds of smoothing members are so varied that in addition to the automatic or non-automatic volume control, a variation of the timbre automatically 55 occurs in accordance with a predetermined law. The means for this purpose are no doubt known. If it is desired to obtain an increase in the percentage of deep sounds, then, for instance, in lowfrequency transformers, the resuit will be obtained that due to the reduction of the general volume and due to the preliminary magnetisation reduced thereby, in a corresponding winding, the inductance of the transformer increases whereby, as is well-known, the transmission of 65 deep sounds is also favoured and vice-versa. If an increase in the anode direct current, or other current component employed for preliminary magnetisation, is to produce an increase instead of decrease in the inductance, then this can be 70 done simply by providing a constant preliminary magnetisation by means of a constant direct current or permanent magnet in addition to the variable magnetisation, in such a manner that the variable and fixed preliminary magnetisa- 75 that variations of the preliminary magnetisation

tions act in opposition to one another. As the variable magnetisation diminishes, the fixed predetermined preliminary magnetisation is then no longer counter-balanced and hence comes into action.

In addition to the methods of automatic volume control and automatic tone control already discussed, it is possible according to the invention to effect tuning of an apparatus quite advantageously, not for the purpose of volume variation, but for the purpose of adjustment to different transmitters. Since it is possible to tune oscillatory circuits merely by varying the direct current intensity of the preliminary magnetisation, this provides, for instance, a method of adjusting a receiver which may be carried out with a very small number of mechanical parts. Since a tuning coil having a Ferrocart or a similar core, and a blocking condenser connected in parallel therewith occupy an extremely small space, a variable tunable oscillatory circuit can in this case be arranged in an extremely small space. The space which this oscillatory circuit requires is so small that it can even be arranged in the base of an electron tube or valve. The variation of the tuning can, in this case, be effected in an extremely simple manner, since it is only necessary to provide on the coil an addltional direct current winding so dimensioned that, with a core of given mass, it allows sufficient variation of the preliminary magnetisation. Since, therefore, there need no longer be any direct mechanical contact between the oscillatory circuit to be adjusted and the tuning means, this method provides an ideal solution for the remote control of receivers. For the operator to tune the oscillatory circuit from any desired distance away, it is only necessary for him to vary the magnetisation by means of a variable resistance or the like, for instance a rotary resistance or else a carbon compression resistance which, as is well-known, allows a smooth variation of resistance. This manner of varying inductance is very favourable also for the arrangement of tuned oscillatory circuit at place where a variation of tuning can be mechanically effected only with difficulty, such as, for instance, within the vacuum space of an electron tube, since the oscillatory circuit constants are varied without mechanical means. This method of tuning is particuiarly favourable in high-frequency amplifiers which work, for instance, with a choke-coil coupling. In such amplifiers, it is preferable, in order to reduce stray capacity, to arrange the coupling choke coil in the interior space of the tube. Thus, for instance, in muitiple tubes containing several systems coupled together, the coupling choke may be arranged within the tube in which case the wave band to be transmitted can be displaced at will from outside, simply by varying the preliminary magnetisation. However, the method is also useful for the tuning of ordinary series high-frequency ampliflers.

As has already been shown, when using threelimb transformers or similar arrangements which prevent coupling between the tuned and the tuning circuits, it is possible to remove any danger of mutual coupling taking place between amplifier stages when several amplifier stages are simultaneously tuned by connecting the preliminary magnetisation windings of all these amplifler stages in series or in parallel. In this manner therefore, it is easily possible to construct a set with single-knob manipulation. In order

may act uniformly on the individual circuits to vary their tuning, it is possible by displacement of turns or groups of turns to pre-arrange that the individual preliminary magnetisation coil and the individual oscillatory circuit coil match one another as accurately as possible; alternatively, however, the cores per se themselves, which are rather soft can also easily be corrected for balancing purposes by cutting off particles or by deformation. Finally, it may be preferable 10 to make the mass cores conical, whereby adjustment may be effected by adjusting the extent to which the mass cores enter each particular coil. Likewise, the small blocking condensers which serve for completing the oscillatory circuits may 15 either themselves be rendered slightly variable by compression screws in order that adjustment may be effected, or else small variable condensers. similar to the well-known trimming condensers provided on ganged rotary condensers may be 20 connected in parallel with the blocking condensers, to permit exact balancing. The variation of the receiving ranges in this case, may be effected either in the normal manner by the connection and disconnection of groups of turns or 25 else by the parallel connection of larger or smaller blocking condensers.

The construction of tuning variometers is also very much easier to carry out by means of ferromagnetic iron cores. since the coupling can be 30 varied by changing the position of the coils to a much greater extent if the flux is concentrated by ferromagnetic cores. The various attempts to construct receivers which allow of receiving the whole wave band, by simultaneous variation 35 of capacity and inductance, on a single revolution of the interconnected variable tuning elements, therefore, become much more promising here without having to resort to such dimensions of the condensers that strong damping and thus 40 unfavourable reception results.

The method of automatic volume control already mentioned above can, of course, also be combined with the described method of tuning by varying the preliminary magnetisation, in which 45 case very simple apparatus are then obtained in which both tuning and volume control are obtained merely by varying the preliminary magnetisation in different windings. Since very few mechanically moved tuning means are required. 50 this provides ideal apparatus, particularly for remote control, since manual volume control can also be effected by variation of the preliminary magnetisation. An apparatus is thus obtained wherein single-knob tuning of almost any desired 55 number of oscillatory circuit together with both manual and automatic volume control is merely by varying the current Intensity of preliminary magnetisation currents, an operation which can be effected from any number of points located at any distance from the receiving apparatus. Such apparatus are particularly suitable in combination with amplifiers constructed in accordance with British Patent No. 415,079. The use of saturating tubes in the anode circuits as described in the above British Patent considerably simplifies the construction of the main parts of the apparatus and the decoupling arrangements. owing to the intense filtering action of these tubes against mains alternating current and also 70 due to the mutual decoupling action of the individual stages, so that an extremely simple apparatus is obtained. In order to prevent the emission from the cathodes, when using photo-cath-

ent upon the general intensity of illumination at the receiving station, artificial illumination is of course used for this purpose. This can be obtained, for instance, in an apparatus which is provided with indirectly heated tubes having high voltage oathodes for direct connection to direct and alternating current, by using for the production of the artificial illumination that part of the mains voltage, which must be taken up in breaking down the mains voltage to the voltage required for the cathode operation. However, since mains fluctuations are manifested in a disturbing manner in this case by fluctuating disturbance of the photo-cathode current, it is desirable to use light radiators which are very inert thermally, such as for instance, lamps, constructed in the manner of the well-known Nernst lamp. This form of illumination is desirable in all cases where normal lamps cause excessive disturbances in cooperation with photo-cells, such as, for instance, in the reproduction of sound films, etc.

By the combination of the above described tuning and volume control by variation of the preliminary magnetisation with the suggestion last mentioned, which are described in the present Applicant's British Patent above referred to apparatus are obtained which are of extraordinary simplicity and reliability. However, by employing the method of tuning based on a variable preliminary magnetisation, it may be difficult in some circumstances to cover a sufficiently large range of reception without the connection and disconnection of condensers or coll parts. In this case, the range to be received may be subdivided into a relatively large number of wave ranges in the manner described in British Patent No. 409,737, in which case, as has been described in detail in that patent, this subdivision has the further advantage that the balancing of several simultaneously tuned oscillatory circuits is substantially simplified, so that a particularly simple form of single-knob manipulation is thus obtained.

The variation of the preliminary magnetisation can also be employed even to obtain the adaptation of the mains transformer to incidentally occurring mains fluctuations or else to construct a mains transformer which can be employed without change of connections on different mains voltages. To this end the effective inductance of the core is varied by means of rectified current from a secondary winding for instance, the current from the secondary winding supplying the anode current which must, ln any case be rectlfied in such a manner that as the main voltage increases, the secondary voltages decrease, at least in those windings where excess voltages are dangerous, that is, particularly in the heating windings. A fluctuation of the anode voltage is generally not so detrimental, owing to the fact that the grid bias is usually produced at the present day by the voltage drop in a resistance, whereby the grid bias automatically adapts itself to the changed anode voltage and automatic control of the volume can be effected in any case. Hence, a small auxiliary transformer which merely supplies the heating current for the tubes, can be regulated by means of a mains transformer, which supplies the anode voltage and the like. The tuning and regulation becomes very simple, of course, when superheterodyne receivers are employed, particularly of the kind according to British Patent No. 301,498 and the addition odes in the saturating tubes from being depend- 75 thereto No. 313,414, since in this case only the

tuning of a single circuit is to be varied, which of course renders any type of balancing superfluous. Automatic noise suppression in combination with automatic fading compensation can also be obtained in simple circuits even with a pure form of magnetisation curve, since the curve exhibits a mean straight part, which would then serve for maintaining the sound volume within certain limits while the first flattening which ocfor noise suppression while the second flattening which occurs at saturation may be employed for the compensation of fading. In the same way, of course, the constructional examples described here primarily for fading compensation can be 15 moving-coil direct current instruments. used accordingly for the correction of amplitude in sound film and sound disc recording and such

The method of inductive tuning herein described may be used also in the case of capacitative tuning, wherein a tuning capacity is used which is varied by a bias. To this end use may be made as tuning condenser of an arrangement such as an electrostatic loud-speaker. wherein by varying the bias the capacity is varied. The above description as applied to inductive influencing may then be read as applying to capacitative influencing.

A number of further possibilities which arise when using mass cores of the nature of Ferrocart material may be mentioned. In the first place, it is possible to obtain mechanical attraction of such cores as with normal alternating current magnets. This fact enables for instance, high-frequency relays to be constructed for use in transmitting and receiving circuit, particularly in receiving circults, wherein a mass core attracts an armature of an identical ferromagnetic high-frequency mass. However, telephone receivers, loud-speakers, sound recording instruments and the like can then also be directly constructed, which enable direct conversion of highfrequency oscillations into mechanical oscillations and vice versa. Similarly voltmeters and ammeters adapted to be directly operated with high frequency, galvanometers, power meters, etc., may be constructed in which, however, the normal types of iron are replaced with such mass material. In this case, amplification of the action may be effected by introducing back-couplings, by means of discharge tubes or other amplifying means. Therefore, for instance, in a telephone, the mass core of which is directly energized by high-frequency currents, amplification can be obtained by the use of high-frequency back-coupling. Of course, low-frequency back-coupling would also be possible per se. Nevertheless, the high frequency back-coupling in this case has the advantage that no appreciable distortion is introduced thereby. Since, however, as has been shown above, tuning may be effected by varying the preliminary magnetisation or the coil-spacings with respect to one another or to the iron core or the like, it is possible in this manner to combine both the tuning circuit and the reproducing instrument and possibly also a volume control all into one constructional element.

In the devices mentioned for the conversion of mechanical into electrical energy, it is possible, 70 for instance, with sensing members, either to vary the coupling at more or less constant frequency by variation of the preliminary magnetisation or the like, or else to vary the frequency whereby the transfer of energy between two cir- 75 the remote controlling apparatus in cooperation

cuits is then varied due to the detuning between two circuits. The above suggestion for the construction of high-frequency instruments operating directly in high frequency by the method usual for low-frequency is also of particular application in directional reception, since in such apparatus the goniometers can then be constructed in like manner to soft-iron instruments. Thus, by more or less intense preliminary magcurs at the beginning of the curve may be used 10 netisation, which can possibly be obtained by permanent magnetic fields, assymmetry of the magnetic fields may be obtained on change of direction of the high-frequency current, so that instruments can even be produced on the system of

By virtue of the asymmetry of the magnetic fields in the reversal of polarity of the current in a sufficiently pre-magnetised iron core, demodulations and modulations can also be effected with such mass cores. Therefore, such a mass core with corresponding windings and pre-magnetising coil etc., may also act as a demodulator in receiving circuits and thus fulfil the function of detector tubes. Since, for instance, in heterodyne receivers, two tubes acting as detectors are required then in such a case by means of the use of suitable coil or transformers, two tubes may be replaced by a mass core. This constitutes a considerable simplification. In order to replace the first tube, such an arrangement would have to act both as a modulator and dcmodulator, but in the second case, for the conversion of the intermediate frequency into the low frequency merely as a demodulator. Another tube is definitely required for the local generation of oscillations in heterodyne reception. Since, however, for the generation of oscillations work can be done on the straight part of the characteristic of a tube, then a low-frequency tube, preferably the output tube, in reflex connection, may be used for the generation of oscillations since in this case, owing to the great flexibility of control complications should least be expected. If the reflex system is also used elsewhere in such receivers, heterodyne receivers may be constructed with very few members. Since the tuning in such receivers can also be obtained by varying the effective inductance in accordance with the aspects given above, a 50 receiver may be designed of a simplicity hitherto not known.

For the remote operation of receivers, it is important that the change-over of the individual wave bands should be effected with the same current as effects the variation in tuning. This can be effected in a simple manner by interrupting and reinserting the current employed for varying the inductance or for volume control, the interruption and re-insertion electro-magnetically setting into motion in the receiver an escapement wheel which sets a wave-band changing shaft into rotation. Since, as above described the actual tuning circuit may be completely decoupled from the circuit effecting the preliminary magnetisation it is immaterial in which direction the preliminary magnetisation current flows. By means of a polarised driving mechanism for the escapement wheel the latter may be rotated. for instance, clock-wise, on the passage of a current in one direction, while on reversal of the direction of the current the escapement wheel is moved in the anti-clockwise direction as the circuit is broken and remade. Therefore, by means of a suitable commutator switch arrangement in

with the escapement wheel mechanism described, it is possible to effect rotation of the wave band change-over switch in the actual receiving apparatus in either direction from the remote controlling apparatus.

Since, the direct current component of the anode current in the demodulation devices owing to the selectivity of the receiving circuits, varies with variation of the tuning when a given transmitter is being received, this current variation 10 may also be employed for the subsequent automatic correction of the tuning, since the direct current component of the anode current can then be used to vary the preliminary magnetisation to effect correction of the tuning circuits. Hence, it is possible in this manner to construct automatic self-tuning apparatus, or apparatus which, once tuning is effected, automatically adapt themselves to variations of frequency of the oscillations received.

Since the inductances provided with mass cores herein described may be used to replace modulation tubes, they also can be used in arrangements made in accordance with a prior invention of the present applicant to replace discharge tubes by inductance coils or transformers. In the arrangement for multiplex telegraphy according to the prior invention use is made of two parallel tubes, the grids of which are energised in opposite phase by a local oscillation. These 30 tubes can be replaced by corresponding transformers of the type herein described. In the prior arrangement relays are shown connected in the anode circuit of the tube. These can now be dispensed with, since the mass cores of the 35 said transformers may themselves be designed as relays and so permit conversion of the oscillatory energy into mechanical energy directly. arrangement according to the prior invention may also be used for the analysis of frequency 40 mixtures, by applying a control frequency variable in accordance with a well-known law to the control grids, whereupon a pointer or the like connected to the relay armature employed therein gives a strong deflection when the control fre- 45 quency conforms with the frequency of a component of the current mixture supplies to the anode circuit, by application of the principle herein discussed, the tubes may be replaced by mass cores provided with a suitable winding and 50 an instrument can therefore be constructed which acts directly as frequency analyser without the aid of discharge tubes, if a known variable auxiliary frequency is supplied. The different constructional examples indicated in the 55 above application can be combined among themselves, so that an extremely large number of possibilities of use is obtained.

A further application of the invention relates to heterodyne receivers and more particularly 60 those which are constructed in accordance with British Patents Nos: 301,498 and 313,414, wherein an aperiodic input or, at least, an input which acts aperiodically over a part of the receiving range is employed as far as the mixing tube, this 63 aspect of the invention however, being in no way limited to those special types of receivers. In such receivers, one of the most important drawbacks is the strong crackling often complained of, which arises from the mixing tube and is 70 then amplified to large amplitudes by the subsequent, usually powerful, intermediate frequency amplification. This amplification following the frequency changer was previously desired when difficulties still existed in the amplification of 75

short waves. At the present day, these difficulties are obviated to a great extent, and it is therefore now proposed to arrange the entire amplification before the mixing tube whereby the noises arising in the mixing tube can be rendered small in relation to the receiving oscillations which reach this tube already fully amplified: In this case, of course, it is immaterial whether the amplification is effected by discharge tubes or by a back-coupling effect.

In addition to the suppression of the crackling tendency, however, other considerable advantages are also obtained by means of this arrangement. For instance, a finer anode rectification can easily be used, since the amplitude of the receiving oscillations applied to the control grid of the mixing tube after amplification can in any case be made so large that work is no longer done on the changing curvature at the lower bend of the characteristic, but the half-waves to be rectified extend into the straight part of the characteristic, so that a linear instead of an exponential action results, which is desirable for many cases. However, it is also advantageous in this receiver to employ the recent mixing arrangements employing pentagrid or hexode tubes such as described in British Patent No: 409,756, with multiplicative instead of the additive or subtractive intermediate frequency formation. Since, in accordance with the present proposal it is important to avoid any considerable amplification after the mixing tube, it is also desirable to avoid subsequent powerful low-frequency amplification. It may therefore be preferable to apply the output from the rectifier directly to the loud-speaker or even, as described above to dispense with all low-frequency parts and allow the high frequency to act directly on an electrostatic or electromagnetic high frequency loudspeaker (provided, for instance, with Ferrocartlike mass cores. Between the rectifler or highfrequency loud-speaker and the mixing tube in place of the usual intermediate frequency amplifler, merely an intermediate frequency filter chiefly comprising several members should be arranged, which may be constructed in a manner well-known per se as a chain conductor or may comprise mechanical resonators such as quartz crystals.

Particularly if, aperiodic or semi-aperiodic input circuits are employed as suggested above, it is desirable to use for the amplification effected before the mixing tube, tubes or combinations thereof according to British Patent No. 415,079, since these tube combinations provide very high amplifications in combination with relatively low internal tube resistances. This is an important point in aperiodic or semi-aperiodic amplifiers owing to the relatively low dynamic resistances which may be given to such circuits for the adaptation thereof to the tube resistances. As has been mentioned in the last mentioned patent, in the case of high-frequency amplification, such series connected tubes must be arranged so that the capacity of the connecting leads of the successive systems, is as low as possible, and to this end tubes with indirectly heated cathodes. which eliminate the capacity of the heating current source, and multiple tubes, which keep the capacity of the connections between the individual systems low, are employed. If several systems are used in this manner, the construction of a multiple tube may become rather complicated and thereby give rise to a large quantity of waste. In this case therefore, instead of com-

bining the whole in a single multiple tube, it is desirable to provide each system or small group of systems in a separate tube, but to construct the individual tubes so that they can be connected one to another with supply leads led out directly instead of through a base, so that in practice, the advantages of the short direct connections provided in multiple-tubes are obtained but in manufacture the advantages of the production of single or two-system tubes exist, the tubes being 10 combined at all into multiple-system units.

Since, when using a comparatively large number of series-connected systems, very high voltage amplification is obtained and also since an anode current source, i. e. say, the mains-trans- 15 former etc., is designed for operation on high voltage and relatively small current, this type of apparatus is particularly suitable for combination with electrostatic loud-speakers.

In these methods of reception with aperiodic or 20 semi-aperiodic input circuits, there is the advantage already emphasised in the prior patents referred to that single-knob manipulation is thereby obtainable in a very simple manner. It is very desirable for the constitution of these cir- 25 cuits to use coils with ferromagnetic cores at least for the construction of the multi-element filters in order to avoid stray fields and in consequence undesirable couplings, and if this is done it is possible to go a step further and, by the use of an 30 adjustable preliminary magnetisation to use these ferromagnetic cores in the manner above described in place of rotary condensers, for tuning in to various waves to be received. The difficulties with this method of tuning, i. e. by preliminary magnetisation, in the case of a plurality of tuned circuits arise from the necessity of keeping the individual circuits in resonance over the entire receiving range, and these difficulties are avoided here, since when using aperiodic input 40 circuits, only one continuously variable tuning means is provided and when using semi-aperiodic input circuits, the necessity of observing absolute accuracy of the tuning between the individual circuits is dispensed with.

The necessity for accuracy of tuning can be still further reduced in the latter case by dividing the total wave range into a large number of small ranges according to British Patent No. 409,737. In particular, if several simultaneously 50 tunable circuits with variable permeability tuning are provided, this is of advantage. With this type of tuning , which is particularly suitable for the remote control tuning of apparatus, the design of the tuning scales at the receiving apparatus or at the individual remote-operating place can then also be very conveniently designed. Since in this method of tuning, as already described in British Patent No. 409,737 the delicate mechanical adjustment is replaced by a form of electrical vernier, a very simple mechanical equipment is sufficient at the individual remote-operating places.

The tuning scales are preferably constructed as "panel scales". At the front of the receiving or remote control apparatus there is provided a plate of, for instance, opal glass of suitable dimensions (say 10 by 20 cms) which is divided into horizontal and vertical strips. If the plate is placed crosswise, strips are obtained which 70 owing to their elongated form, are of suitable shapes for the inscription of station names. Each column then corresponds to one of the relatively small wave bands and the individual strips constituting a column correspond to the indl- 75 just discussed permit in other respects the use

vidual stations to be received within this band. The scale may be illuminated from the rear and there may be arranged on the apparatus as a wave switch, a plane sliding switch, the switching arm of which is movable behind the scale in such a manner that when connection is made for the reception of the shortest wave band, it is situated just behind the horizontal row which corresponds to the shortest band and, when the wave band connection is changed it comes into the corresponding position behind the wave band now switched in. If a small lamp is fixed to the switching arm, the range switched in at any time is always illuminated. By means of the knob effecting the tuning within a selected range, the lamp may then be moved to and from in a vertical direction behind the scale plate on a runner or the like, so that the strip bearing the name of the station being received is then illuminated. Alternatively a long horizontal pointer may be provided which moves in front of or behind the scale and which indicates the station in the column illuminated at the time, to which the apparatus is tuned. Such a panel scale, as will be hereinafter described can also be suitably constructed for use at a remote-controlling post.

Remote tone control by the variable induction of pre-saturated low-frequency chokes or transformers may also be employed.

The remote tuning mechanisms herein described are, of course, not limited to the provision of a panel scale as herein described any more than the panel scale is limited to this manner of application, since it is suitable generally on sub-division into a larger number of individual ranges, particularly in the arrangements according to the prior British patent No. 409,737 above mentioned. The methods of execution of the limited tuning or operation herein described are not only applicable to the special methods of reception described here and in the last mentioned patent although they are particularly suitable therefor.

The method according to the invention, can  $_{45}$  be used with very great advantage for yet another element in the tuning of a set, namely, not for operation of a continuously variable tuning element i. e. for replacement of the rotary condensers, but may also be used to replace the stepby-step tuning of an inductance by wave change switches. In this case, special advantages are obtained. As is well-known, it is a great disadvantage of all apparatus working with a wave change that change-over switch contacts must be provided in the high-frequency circuits, since. firstly, the connection of the lines from and to the contacts causes difficulties, since all capacitative effects in such circuits are to be avoided to the greatest possible extent. In addition, the lines to the change-over switches and the changeover switch contacts give rise to additional dampings, apart from the undesired capacitative couplings, and finally also the mechanical structure of these multiple switches is complicated and expensive.

All this can be avoided by the step-by-step switching or pre-magnetising currents, which offers great advantages, in particular, when a large number of changes of connection is to be effected, such as, for instance, in an apparatus constructed according to British patent No. 409.737.

All tunings effected by saturation such as hereinbefore mentioned and more particularly, those

of much higher permeabilities than, for instance, are obtained with the Ferrocart-like materials wherein during operation, the inductance value remains unchanged. This is due to the fact that in the case of the materials the permeability of which remains unchanged, the permeabilities must be so chosen, that even at the smallest wave length arising, the damping is not too great, whereas in carrying out the present invention, the permeability is so chosen that in the condi- 10 by-step alteration of an inductance, tion of the maximum preliminary magnetisation, the permeability is a minimum, the damping is kept within the desired normal limits for the smallest wave lengths, while in the case of higher wave lengths at which a higher permeability does 15 not bring any objectionable increase in damping, this increased permeability can then also be effectively utilised. This applies particularly to the step-by-step inductance switching here described, since here the maximum permeability 20 comes into action at wave lengths which behave particularly favourably as regards the damping with ferromagnetic cores.

The step-by-step tuning by varying the current intensity or voltage of a pre-magnetising current 25 in steps may even be used in the case, owing to its special advantages, in arrangements which are otherwise normally tuned, that is tuned by rotary condensers or variometers or in combination with the methods of tuning hereinbefore 30 described, so that the step-by-step and also continuous tuning can then be effected by the same or even different preliminary magnetisation coils on the same or on different cores.

It is also very suitable, particularly for remote 35 tuning, however, to combine the continuous tuning (described above) with a step-by-step tuning by variable preliminary magnetisation by means of electric biasing of condensers, for instance Seignette salt condensers, for the purpose of 40 remote operation of receiving installations havlng several tuning circuits by a single-knob.

In particular, for such cases in which a high degree of curvature of the magnetisation curve is desired in addition to a high degree of varia- 45 tion of the permeability at the working point, that is, especially in those cases where a modulation phenomenon is desired, there is a certain drawback, with cores such as Ferrocart cores, having a fine subdivision in the distribution of 50 direction of the lines of force, the magnetisation curve presents only slight curvatures. According to a feature of the present invention therefore, a magnetic core is provided which is hereinafter described, and which has proved to be particu- 55 larly suitable in such cases, and also in all the other cases which are mentioned in this application.

Since very sharp bands can be obtained in the magnetisation curve even with relatively small 60 magnetomotive forces, if the ferromagnetic material is suitably chosen, this material is also particularly suitable for the construction of magnetic amplifiers which operate on the principles well-known per se, such amplifiers could only be 65 constructed for relatively low frequencies in the past, owing to the high iron losses.

In order that the various aspects of the invention may be more clearly understood, a number of embodiments thereof will now be described 70 with reference to the accompanying drawings in which

Fig. 1 shows a simple volume control arrange-

Fig. 2 shows a modified form of volume con- 75 ed above, additional volume control may be ob-

trol arrangement similar to that shown in Fig. 1, Fig. 3 shows an arrangement for obtaining automatic volume control.

Fig. 4 is still another form of volume control, Fig. 5 is a section of a form of core suitable for use in carrying into effect the various features of the invention.

Fig. 6 is an elevation,

Fig. 7 shows an arrangement for effecting step-

Fig. 8 shows diagrammatically the circuit arrangement of a superheterodyne radio receiver embodying a feature of the invention,

Fig. 9 shows a tuning indicator arrangement and remote controlling apparatus for use in conjunction with the invention.

Fig. 10 is a remote control circuit which may be employed, according to the invention, in eonjunction with the other remote control arrangements described.

Fig. 11 shows a reactor having the high frequency core arranged between the poles of an energising magnet,

Fig. 12 shows the invention applied to capacitative tuning,

Fig. 13 shows a modification of Fig. 12,

Fig. 14 shows a circuit for obtaining automatic tuning control,

Fig. 15 shows a more detailed circuit for obtaining automatic tuning,

Fig. 16 shows another circuit for obtaining automatic tuning.

Fig. 17 shows a modification of Fig. 16 in which the automatic tuning is effected by means of a variable capacity,

Fig. 18 shows a push-button remote control cir-

Fig. 19 shows a remote control circuit with a sub-divided wave range,

Fig. 20 shows the lay-out of a control panel.

Referring first to Fig. 1, which shows a volume control arrangement a tube I is shown adapted to operate, for instance as an anode current rectifier, so that an increased amplitude at the grid, produces an increase in the anode direct current component. An oscillatory circuit 2 is connected to the grid of the tube. The anode circuit of the tube is provided with a back-coupling coil 3 and a variable leak condenser 4. According to the invention, the back-coupling coil is coupled to the oscillatory circuit through a ferromagnetic core 5. the inductance of the ferro-magnetic core being of such a value that favourable operation of the back-coupling is just obtained with the preliminary direct current magnetisation due to the anode current which passes at the desired volume. The adjustment of such a close back-coupling is effected by means of a variable condenser 4. If the capacity of the condenser 4 is made very large the high frequency component is mainly conducted to the cathode without traversing the coil 3 and a slight back-coupling action takes place, while on the other hand, if the capacity is reduced, the back-coupling effect increases. If strong signals are now received, which cause a substantial increase in th anode current, the action of the back-coupling decreases accordingly since the increased preliminary magnetisation reduces the mutual inductance of the coil 2 and The reverse action may also take place. In this case, therefore, a very simple type of volume control is obtained. Any re-tuning of the circuit 2 which occurs may be compensated in any suitable manner if it is undesirable, although as stat-

tained thereby. Any desired adjustment of the time constant of the regulating operation can be obtained by condensers and resistances or the like, suitably incorporated in the anode circuit.

In Fig. 2 a modification of regulating device for 5 the regulation of back-coupling is illustrated, which has an even more intense action to some extent. In this case, a three-limb core of ferromagnetic material is provided, which carries three different coils or groups of coils. The coils 10 1, 2 are connected in a tunable oscillatory circuit, which as in the Fig. 1 arrangement, may be connected to a tube acting as rectifier. These two coils are wound in such a manner that the field they generate in the coil core are in series. The 15 field produced by them, therefore, does not substantially penetrate the centre limb of the transformer. On the centre limb of the core there is arranged a coil which is traversed by the highfrequency components of the anode current of 20 any succeeding high-frequency amplifier or the detector tube to which the oscillatory circuit 1, 2 may be connected. As long as the inductance in the two symmetrical portions of the three-limb transformer I and II is uniform, no back-cou- 25 pling effect will occur at all, since the oscillatory circuit is decoupled from the back-coupling coil in this condition. The core of ferromagnetic material or the oscillatory circuit coil may be so designed that complete mutual decoupling occurs when a 30 certain normal current flows through the turns 3. 4 connected in opposition to one another which are traversed by the direct current component of the de-modulation tube. If the volume falls below this desired degree, then the consequent re- 35 duction in the preliminary magnetisation by the two coils, 3, 4 unbalances the symmetry in the two magnetic circuits I, II so that coupling takes place between the oscillatory circuit 1, 2 and the back-coupling coil, such coupling being intended 40 to act in such a manner that the back-coupling has an amplifying action. If, however, the field strength of the incoming oscillations becomes very great, so that the anode current in the rectifier tube considerably increases, the symmetry of 45 the arrangement is displaced in the opposite direction and the back-coupling coil now acts as additional damping, that is reduces the volume. For this effect to be produced, of course, it is necessary that initially, that is without the action 50 of the coils 3, 4 asymmetry exists between the circuits 1, 2 so that a close back-coupling is then already present, this asymmetry being gradually removed as the direct current in the coils 3, 4 increases, extending increase in the current in the 55 coils 3, 4 leading to asymmetry in the opposite direction. The arrangement just described owing to mutual compensation of the action in the colls 1, 2 provides automatic volume control without any variation of the tuning. A third example of 60 an arrangement providing automatic volume control is shown in Fig. 3. In this case, the coupling per se is varied in place of the back-coupling. I is, in this case, the input circuit of a coupling arrangement, while II is the output circuit. The 65 arrangement may be a transformer coupling two high-frequency amplifier stages or the like. As can be seen, the transformer has 3 groups of windings, namely, a coil I which is connected to the input circuit, a centre-tapped coil 2 from 70 which the output circult is so derived that, if the coil is tapped symmetrically no oscillatory energy can reach the input circuit and, furthermore, a second two part winding 3, 4 through which the

tube flows. In this case also the tapping of the coil 2 may again be made asymmetrical at the outset or the mass core may be asymmetrically designed or both, so that normally, without direct current in the windings 3, 4 or with only a small direct current component, a considerable asymmetry is present, whereby an essential part of the input at I may be derived at II, while as the preliminary magnetisation by the coils 3, 4 increases the asymmetry is gradually removed and finally the bridge or the differential system. on very strong preliminary magnetisation, gradually approaches balance, so that the current component which is transferred from I to II continuously diminishes.

If the coils I and 2 are parts of tuned coupling transformers between individual amplifier stages, where the detuning may be undesirable. this de-tuning may be counterbalanced as has already been described in connection with the example shown in Fig. 2 by the counterconnection of similar elements so that no undesired de-tuning of the circuit then occurs.

The great advantage of such volume control arrangement resides in the fact that although no special tubes and no special grids in the tubes are employed, a variation of the volume can be obtained which, in muiti-stage apparatus, such as multi-stage amplifiers, where a regulating stage may be arranged before each amplifying tube, multiplies the action in the individual tubes, and so allows of obtaining regulation within very wide limits. In the arrangement described in Fig. 3 wherein the preliminary magnetisation only acts to vary the energy transmitted from circuit I to circuit II it is also possible since there are no tuning problems in this case to effect the regulation in the low-frequency part or else in the high-frequency and low-frequency part together. If the regulation is effected in the low-frequency part, it is possible to dispense with the use of special types of iron.

A further development of the invention idea is shown in Fig. 4. In this case, I is against the input circuit of a coupling arrangement which is variable for the purpose of volume control and II the output circuit. I and 2 are normal fixed inductance coils, while 3 and 4 represent two inductances variable by preliminary magnetisation. The bridge arrangement comprising the four resistances 1, 2, 3, 4 is initially balanced, that is, when only the preliminary direct current magnetisation by the battery 5 is effective in the coils 3, 4 and there is no preliminary magnetisation by the centre tapped coils 6 and 7, then the bridge is in equilibrium. If preliminary magnetisation arises in the coils 6 and 7 the equilibrium of the bridge is disturbed, since the inductance of the coil on the core 3 is increased and the inductance in the coil 4 is decreased to equal extents. A considerable flow of current will then take place in the arrangement from the input part I to the output part II. The preliminary magnetisation of the cores 3 and 4 is effected in the following manner. The two tubes 3 and 9 again represent rectifiers, which are energised by an oscillatory circuit 10, which may be assumed to be connected to the output of a high frequency amplifier. The circuits I and II may be the coupling between two or more stages within this high-frequency amplifier. The tubes 8 and 8 then act as rectifiers and in the present figures, for the sake of clearness, the part of the lines which leads to the low-frequency amplifier is direct current component of the demodulation 75 omitted and only that part of the lines of the an-

ode circuit of the tubes 8 and 9 is shown, which carries direct current. In this case, means not shown, may also be provided which effect the filtering of the alternating current component.

The two tubes 8 and 9 are intended to work 5 with somewhat different blases, e. g. tube 8 with normal negative bias, so that it acts as a normal anode rectifier tube, and the tube 9 with so great a negative bias that the tube begins to function only at considerable amplitudes. When signals 10 of very low intensity are received, so that it is to be feared that the material part of the audion only comprises strongly reproduced disturbances, then owing to the small anode direct current of the tube 8 and the tube 9 in the eoils 6 and 7, no 15 appreciable direct current at all will flow, so that the original equilibrium of the bridge is not disturbed and practically no output is transmitted from I to II, that is, not even the intense disturbing noises, which are otherwise always pres- 20 ent as a disturbing factor in a sensitive receiver on adjustment to maximum volume. If signals of a moderate sound intensity occur, the anode direct current of the tube 8 increases and, hence an increasing magnetisation is produced in a half 25 of each of the coils 6 and 7, which as has already been mentioned, act in the opposite sense on the cores 3 and 4 so that in one core the preliminary magnetisation proceeding from the battery 5 is increased and in the other, it is decreased. Thus 30 the equilibrium of the bridge is disturbed and a good transmission of the energy from the input circuit from I and II is effected. If the signal intensity considerably increases the tube 9 will also come into action, so that if the signal am- 35 plitude becomes greater than the negative bias of the tube 9, this tube then also passes an anode direct current. Hence anode current is now supplied to coils 6 and 7 from both tubes 8 and 9. Since the two halves of the coils 6 and 7 act in 40 opposition to one another, however, simultaneous operation of the tubes 8 and 9 acts to remove the preliminary magnetisation in the coils 6 and 7 and the bridge is again balanced so that the amount of energy which is supplied from the in-  $_{45}$ put circuit I to the output circuit  $\Pi$  is kept constant.

Here again therefore an arrangement is provided, wherein automatic volume control and automatic noise suppression is provided. coil combination shown here can also be arranged in several stages of a multi-stage high-frequency amplifier or even a low-frequency amplifier, so that regulation within very wide limits can take input circuit I can be tuned, since the total inductance of this circuit (if no considerable load is applied to circuit II, that is, for instance, only the input of a tube amplifier) does not vary. The total inductance consists of two part-inductances connected in parallel, i. e. the partinductance of the coil in series connection with the iron-core coil 3 and, the part inductance of the coil 2 in series with the iron core inductance Since, however, in the regulating operation, 65 the inductances of the coils 3 and 4 are varied in opposite sense, then provided this variation takes place to some extent in a linear manner, as may be obtained within wide limits of time by a suitable coil and core arrangement for instance, coni- 70 cal iron cores or the like may be used, the total inductance will not vary during the regulating operation, so that these coils may be parts of a tuning circuit.

heterodyne radio receiver constructed according to a feature of the invention. As is readily apparent from the drawing, the apparatus comprises two stages of series connected amplifier tubes, each comprising two tubes 21, 22 and 23, 24 connected in the manner according to British Patent No. 415.079 the two stages are coupled through a band-pass tuned circuit 25. The output of the two stages of high frequency amplification is applied through a further band-pass tuned circuit 16 to the mixer tube 27 shown as a hexode or four-grid tube. The intermediate frequency from the mixer tube 21 is applied by way of transformer 28 and filter 29 to a second detector 30 in the anode circuit of which the loudspeaker L is connected.

The high-frequency signals picked up by the aerial 31 are amplified in the two stages of high frequency amplification to such an extent that. after frequency changing by the tube 27 and detection by the tube 30, a sufficient output is derived to drive the loudspeaker L without lowfrequency amplification.

Preferably, the tuning arrangements are constructed according to British Patent No. 409,737, whereby the condensers forming part of the coupling circuits 25 and 28 and the tuning condenser for the local oscillation generating circuit 32 may all be ganged together without difficulty and may even be condensers of the solid dielectric type.

In Fig. 9 a further development of the invention is shown. In this figure, T is an instrument which is traversed by the pre-magnetising current and indicates its strength. However, since this current is proportional to the receiving frequency, the instrument can be directly calibrated in wave-lengths. By means of a second lever pivoted at D, a parallel ruler mechanism is formed whereby the long pointer Z may be maintained horizontal and guided up and down in front of or behind a panel scale divided, for example into five parts. Connected in series with the instrument T is a regulating resistance R, which enables the intensity of the pre-magnetising current and thus the tuning of the receiving apparatus to be effected. The resistance R is operated by a knob K which is mounted on a long spindle, which may be of square or other crosssection so that on rotation of the knob, the contact arm of the resistance regulator R is turned, but on movement of the knob in the direction of the spindle, the latter may slide freely.

The spindle A may therefore be used to displace. In this case, it is noteworthy that the  $_{55}$  place the contact Y on the slide S by push or pull, so that it can be brought into engagement successively with the contacts i to 9. This contacting causes circuit interruptions or closures which cause the rotation of an escapement wheel in the receiving apparatus in the manner described above, said wheel causing the wave range change-over in the apparatus.

In this case, pressure on the knob K in the axial direction, moves the member M sliding frictionally on the spindle to effect a contact closure between the elements of the controlling apparatus and the lead U, while a pull on knob K, moves the member M in the reverse direction and makes contact between these elements and the lead V. The individual current impulses produced by contact of the contact Y with the individual contact points cause rotation of the escapement wheel effecting the wave range change in the apparatus in either one or the other direction. Referring now to Fig. 8 this shows a super- 75 according to whether the impulses are transmitted to the receiving apparatus through the lead U or V, since the two lines U and V are connected either to different lifting magnets or drive a polarised escapement wheel mechanism in either direction. The result is thus obtained that the position of the escapement wheel corresponds at any time to the position of the contact Y.

On the contact Y there is arranged a small lamp La which illuminates the area below which the contact for the time being in operation is lo- 10 the operation at each point independently. cated. The arrangement is preferably so devised that by means of vertical screens or the like, the adjacent fields are prevented from being noticeably illuminated, only the one area, corresponding to the wave band in use being illus- 15 trated, and the shadow of the long transverse pointer Z being seen on the opal glass or like plate only at that locality. The sixth contact point is shown in the present constructional example as an idle contact, so that by shifting the 20 knob K into its end position, the apparatus can be switched off by interrupting the current interruption. It is preferable to ensure, by means of resilient stops or the like that the contact cannot remain in an intermediate position between 25 two contact points. Furthermore, a change-over to gramophone reproduction, for instance, at the receiving apparatus may be distinguished at the remote-control apparatus by reversal of the direction of the current or the like or by means 30 of a special signalling lamp or by deflection of the instrument in the opposite direction, whereby the pointer is moved out of the range of the tuning scale and replaced by a template with an inscription which is depicted on the scale.

If remote control is to be effected from more than one point, the remote control line may be permanently connected up to the several points (if a movable wire is not employed) and plug connections may be provided at these points into 40 which a remote control apparatus of the abovedescribed kind can be plugged. It is then only necessary, for instance in a dwelling-house with several living rooms from which the operation is to be effected at will, to carry about the small 45 and light remote-control apparatus to the room from which control is desired. Alternatively a remote control apparatus may be permanently installed at each point which shows on permanently connected ammeters or voltmeters, 50 which are calibrated in wavelengths in the above described manner, the wavelength adjusted at the time from that or another place. In the latter case, however, it is preferable to provide a locking arrangement, whereby an adjustment 55 cannot be made from one point if the apparatus is already being controlled from another point, since otherwise the escapement wheel mechanism in the apparatus might fall out of step. The connection and disconnection of the apparatus, however can be effected in this case from any number of points by means of a circuit similar to the "series circuit" in lighting installations.

The individual operating points can be made more dependent of one another by the following arrangement. Means such as a holding magnet as used in the selectors for automatic telephones may be provided arranged on a device which may resemble the above-described remote control, which holding magnet, retains the knob K by 70 means of a braking action of a pawl and stops, in each position into which it is placed by means of the adjuster, in opposition to a restoring force formed e. g. by two springs. If the current is interrupted, the holding magnet or the holding 75 preliminary magnetisation, for instance the bat-

magnets become currentless and the two springs i. e. a torsion spring and a lifting spring acting as restoring forces, rotate the resistance R into its zero position or initial position in which the circuit is broken, and move the contact Y into the idle contact 6. Since, by the use of a "series circuit" as in illuminating installations, the current interruption can take place at any operating point, this provides a possibility of effecting

In this case, as described above, the reading instruments may remain connected at each point if desired, in order that the station to which the apparatus is adjusted may be observed at each point. If several wave bands are employed, a device must also be provided to indicate the particular wave band. This could be effected, for instance, by the arrangement of a separate signalling lamp under each scale section corresponding to a wave band, the lamps corresponding to each section being connected in series or in some other interconnection with the corresponding lamps at the other operating points. Loudspeakers may be permanently installed at the individual operating points or may be plugged in, in the same manner as the remote control apparatus. Alternatively a loudspeaker may be incorporated in the transportable control apparatus. When the apparatus is to be employed as a gramophone connection the changing of discs will necessitate providing the sound pick-up at the remote points, and talking machine connections may therefore be provided at these remote controlling points, in which case care should be taken in the manner described above that the use of the apparatus for the time being as a talking machine is indicated at the other operating points.

The method of remote control above described, is not, of course, confined to its use for living rooms but may be used also when for any constructional reasons, the apparatus to be operated must be arranged at a place, which is inaccessible to the operator, thus, for instance, on automobiles, aircraft, boats and the like.

In such cases also it may happen that the operation is to be effected either from the pilot's or driver's seat, or alternatively from the back seats, in which case at least two independent operating posts may be effected in the manner already described or else by connecting all the lines concerned to each operating post as in a house telephone line-selector.

A further possible arrangement for independent operating posts is shown diagrammatically in Figure 10 by way of example. In this figure two such posts, are shown, in an installation such as for a motor-car wherein the apparatus is to be controlled from the back seats or from the driver's seat alternatively. The principle of the arrangement consists in using resistance branches, wherein by adjustment of regulating resistances, potentiometers or the like at different controlling posts, voltages or current intensities for the preliminary magnetisation may be derived. By this arrangement adjustment at one controlling post need not be varied or brought into the condition of rest before another controlling post can be employed, this result being achieved by the fact that the current consumers of the regulating resistances or the like which are provided at the various controlling posts are connected in a common circuit. Referring to Fig. 10 B is the current source required for obtaining the

323,450 11

tery of the motor-car. The pre-magnetising winding in the apparatus, should be so arranged that the whole wave band for the time being connected up may be covered by variation of the voltage from zero up to half the available battery voltage. In other respects the windings may be constructed in all respects in accordance with the information set out above. The voltage applied to the pre-magnetising windings may be read off at either of the two controlling posts I and II from a voltmeter VI or V2 calibrated in wave lengths or frequencies or station names, permanently connected in circuit. In the position shown, the sliders of the two regulator tappings Ri and R2 are approximately at the centres of their corresponding resistance strips. In this position they both have the same potential and zero voltage is therefore applied across the winding S connected between them. The longest wave is therfore tuned in. If one slider  ${\bf R}$  is moved out 20of its position of rest, then independently of the position of the other slider, at least half the battery voltage may be applied to the winding S. The fact that, in this arrangement the direction of current may change is immaterial for the 25 effect, since the winding S is completely decoupled. However, it is necessary to use voltmeters V, which give the same pointer deflection whichever the direction of the current, unless opposite deflections are expressly desired for special 30 purpose. The wave range change, however, cannot be effected by means of change of direction of current and polarised escapement wheel mechanism, but must be effected in some other manner, such as has been described above. The con- 35 still better. nection and disconnection may also be effected independently of one another at the control posts by "series-connection". Very often, a volume adjustment is required which may also be effected in the manner described above by means of saturating windings or in the same way as the tuning by electrostatic biasing of condensers dependent on voltage (for instance, Seignette salt dielectric condensers).

Turning now to Figure 7 this figure shows by way of example an embodiment of a further feature of the invention according to which the stepby-step variation of an inductance is effected by step-by-step variation of the preliminary magnetisation applied to a core with which said inductance is provided. In this figure, B is the current source, employed for preliminary magnetisatlon, which is shown here as a battery but which in most cases will be replaced by a connection to the mains. R is a resistance variable in steps, which however, if continuous tuning by saturation is to be obtained in addition may also have a continuously variable regulator connected in series therewith. Alternatively the resistance R may be continuously variable between the steps, and may be provided with stops, so that sub-division into several separate wave bands, even with a large tuning range of an apparatus, for instance from 200 to 2000 metres or more, is then dispensed  $_{65}$ with. K is the ferromagnetic core which carries the relatively de-coupled premagnetising and oscillatory circuit windings. Further oscillatory circuits or back-coupling circuits of the apparatus, not shown here for the sake of simplicity are 70 connected to the terminals X. Finally, Z is an additional resistance regulating arrangement the value of which may be adjusted to balance the initial values pre-magnetisation of the individual

ing windings of the cores are connected in parailel and not in series as shown here, the additional resistances Z should not be arranged in parallel but in series with the pre-magnetising turns.

Figs. 5 and 6 show diagrammatically a form of core particularly suited to the various purposes of the present invention. Referring to these figures II is a coil body of thick pasteboard or other insulating material, which serves not only as a former for the coil, but also to ensure that the winding on the core is sufficiently spaced from the conductive core material, at least, where it is desired to provide a high-frequency winding disturbing capacity couplings through the core material are to a great extent avoided. The coil former, which is shown homogeneous in the drawing may also be in the form of a framework structure, so that the H. F. turns touch the insulating material only at a small number of points, a sufficient air spacing for the reduction of capacitative noises being provided elsewhere. Since the finished core in the form herein described forms a self supporting whole, the insulating former can therefore be mechanically relieved to a great extent.

On the coil former there is wound a layer of thin ferromagnetic wire, for instance, iron wire or better an alloy such as permalloy, in such a manner that a small air space is left between the individual turns, which considerably reduces capacity coupling between the individual wire turns. The thickness of the wire should not greatly exceed a diameter of 0.03 mm. thinner wires being still better.

After a layer of wire has been applied in the manner described, leaving sufficient spacings between the individual wires, this layer is preferably fixed and coated by a lacquer of good insulating qualities and of the lowest possible dielectric constant. Following this is an insulating layer, such for instance as paper, which is also fixed and then again, in the same manner as has just been described, another layer of wire or strip is applied. The wire or strip may possibly be produced electrolytically by deposition on a conductor base which is then dissolved away.

In the drawing 12 denotes the intermediate paper layer, and 13 the individual wires; 14 50 shows the arrangement of a layer of the H. F. coil winding. Since the core, as mentioned above, has sufficient mechanical stability after hardening of the lacquer, a direct current winding which is necessary for obtaining a preliminary magnetisation may be wound directly on the core. Hence, at the point where the direct current winding is arranged, the coil former is preferably removed since no reduction of the capacitative coupling is required, and a highly desirable gain in winding space is obtained.

Owing to the relatively low iron or the like content of the core, the permeability of course becomes considerably less than in solid cores or cores normally made of laminated material, but this is desirable for high frequency purposes owing to the consequent reduction in iron losses to tolerable limits. Moreover, sufficient saturation can be obtained in the matter with less current.

connected to the terminals X. Finally, Z is an additional resistance regulating arrangement the value of which may be adjusted to balance the initial values pre-magnetisation of the individual cores against one another. If the pre-magnetis- 75 points, provision should be made for good mag-

netic contact where the flux enters and leaves the core.

The cores described can be most simply produced economically in the form of closed circular rings since they can then be made on a simple coil winding machine with an arrangement well-known per se for the arrangement of intermediate paper layers between the turns, and it is therefore desirable to use two such hereinbefore described. The direct current winding may then comprise two separate windings each arranged on a core, instead of the single winding illustrated in the figure. This last form of construction should even be the 15 of the received signals, this controlling potential better one in regard to the attainment of a favourable filling factor.

The direct current winding and alternating current winding may generally be interchanged.

Referring to Fig. 11, the reactor comprises an 20 electromagnet 41 which is energised by the en. ergising winding 3. The magnet 41 is constructed of waterial of relatively high permeability, such as from iron laminations or solid iron. It may be made from material known 25 commercially as "Permalloy."

Between the pole pieces of the magnet 41 is arranged an inductance 2 wound upon a ferromagnetic core 5 of low magnetic permeability. This inductance may be constructed as described 30 with reference to Figs. 5 and 6. It is preferable to provide a good magnetic contact between the poles of the magnet and the core.

By varying the current flowing in the winding 3 the field strength between the poles of the 35 magnet 41 is varied whereby the saturation of the core 5 may be varied and thus also the inductance of the coil 2. In this way the reactor provides a simple means of adjusting the inductance of the coil 2 to any desired value with- 40 receiver one of the circuits may be provided with out any moving parts being necessary. The inductance 2 may form part of a high frequency oscillatory circuit of a radio receiver so that by means of the reactor according to this invention it is possible to effect the tuning of the circuit 45 ently of the variation of the second circuit, by merely varying the current flowing through the winding 3. This may be effected from the remote point by simple variation of a potentiometer as illustrated in Fig. 7.

It has furthermore been mentioned above that 50 the statement made for inductive operations when using cores of ferromagnetic material may also be applied to capacitative operations when using condensers dependent on voltage. Suitably constructed Seignette sait condensers are particularly suitable as such condensers.

Fig. 12 shows an embodiment employing such a condenser, the circuit arrangement is substantially identical with that shown in Fig. with the exception that the reactor 5 is replaced by the condenser 50 and the condenser 6 by the inductance 60. R is the resistance, which in this case is arranged as a potentiometer across the battery B, whereby the potential applied to the plates of the condenser 50 may be varied to vary the capacity of the oscillatory circuit comprising the condenser 50 and the inductance 60. The condenser 50 may be of the type which has a plezo-electric dielectric or in which the electrode spacing is a function of the potential dif- 70 ference therebetween.

Fig. 13 is a modification of Fig. 12. The condenser with plates 50, 51 has an inductance coil 53 connected across the plates, said condenser and coil forming an oscillatory circuit. The 75 frequency to which the inductance 120 is tuned.

condenser is connected to a source of continuous current, for example a battery 56 through two choke coils 54, 55, and a resistance 57 capable of variation by means of a sliding contact 58 bridges the battery terminals. Variable voltages are thus available whereby variable potentials can be selectively applied to the two plates 50, 51 of the condenser, causing the capacity thereof to change with the result that the tuncores for carrying out the various connection 10 ing of the oscillatory circuit 50, 51, 53 can be varied.

> By deriving from a receiver a controlling potential dependent upon the frequency of the signals received or the variation of the frequency may be used for automatically tuning the oscillatory circuits of the receiver in the same manner as described above for automatic volume control whereby an automatic correction of the tuning to a desired frequency may be obtained. Since the direct current component of the anode current in demodulating devices, owing to the selectivity of the receiving circuits, varies with variation of the tuning when a given transmitter is being received, then this current variation may be employed for the purpose of automatic tuning. The direct current component of the anode current would be used to vary the preliminary magnetisation to effect correction of the tuning circuits. Any of the arrangements described above for automatically controlling the tuning in dependence upon a controlling current or potential may be used for this purpose. This possibility of automatic tuning correction is of particular advantage in the case of remotely controlled receivers which are particularly simple to construct with the arrangements according to the present invention.

> When two tuned circuits are provided in the automatic tuning correction means whilst the other circuit may be provided merely with manual control means whereby the first circuit is kept in tune with the received signal independwhereby accurate ganging between the two circuits may be dispensed with.

The principles employed for automatic tuning control will be explained with reference to Fig. 14 in which the incoming signals pass through the condenser 115 and the resistance 116. Shunted across the resistance is a rectifying device 117 which is in series with the magnetising winding (16 on the magnet 119. The preliminary 55 magnetisation may be produced by a second winding 1181 arranged on the magnet and fed with current from a battery 161. The energising current may be controlled by the resistance 162 which may be arranged at a remote point and serves manually to adjust the tuning to receive any desired station. Alternatively, the winding 118 may be the only winding on the magnet, and have both the automatic controlling current and the preliminary magnetising current passed through it. Arranged in the gap of the magnet 119 is the inductance coil 120 which forms part of an oscillatory circuit, for example the tuning circuit or the local oscillator circuit in a superheterodyne receiver.

It is known that if the frequency of the incoming signals increases, more energy passes the condenser 115 and thus more current is passed through the winding 118 on the magnet 119. thereby increasing the field and increasing the

whereas if the frequency becomes lower, that is the wave length of the incoming signals increases, the reverse action takes place and the frequency to which the oscillatory circuit is tuned also decreases.

The circuit illustrated in Fig. 14 is given for the purpose of explanation and it is assumed that the incoming signals from different stations have the same energy. In practice, of course, the control would be effected through a device for producing equal amplification of the incoming signals, such as for example the last stage of an automatic volume control circuit, or a voltage or current limiting device should precede the amplifier.

Fig. 15 shows a more detailed diagram in which the output from the valve 121 passes through the condenser 122 and the resistance 123, the current in which varies with the frequency due to the variation in energy which passes through the condenser 121 with frequency changes. The current passing through this circuit is rectified by the rectifier 124 and fed to the magnetising winding 125 of the permanent magnet which automatically adjusts the tuning of the input or local oscillator circuit 120.

Fig. 16 shows an alternative circuit diagram for automatic tuning control in which the control is effected by balanced circuits. The principle is illustrated in the drawing as being applied to a superheterodyne receiver. The intermediate frequency signals are fed to the grid of the valve and the signals in the inductance 127 in the anode circuit oscillate at the same frequency. Coupled to this inductance are two tuned circuits I and II, the circuit I being tuned to a wavelength slightly below that which is to pass the intermediate frequency stage and the circuit II to a wavelength slightly above that to be passed. The outputs from these two circuits 40 are fed through the rectifier 126, the cathode of which is connected to the mid-point of a winding 129 on the permanent magnet. The opposite ends of this winding 129 are respectively connected to the circuits I and II as shown. This 45 winding 129 may be arranged on the magnet which has a second winding 130 which produces the field which may be varied to adjust the tuning to different transmitters. In the gap of this magnet is positioned the inductance which 50 is in the local oscillator circuit 131.

The operation of the device is as follows.

If the intermediate frequency slightly increases the circuit I would be more energised than circuit II thus upsetting the balance in the induc- 55 tance 129 and causing more current to pass through the half 129a than the half 129b, which thus has the effect of altering the local oscillator frequency to retune the circuit. On the contrary if the intermediate frequency becomes slightly lower the reverse effect takes place due to the circuit II being more strongly energised than circuit I.

Fig. 17 shows a modification of Fig. 16 in which the rectified output from the two circuits I and II is used to control the frequency of the local oscillator circuit by varying its capacity. To this end the outputs from the circuits I and II vary the potentials produced in resistances 132, 133, which are in circuit with the control electrodes 134 which adjust the capacity of the condenser 135 in accordance with the potential impressed upon those control electrodes. The battery 163 and potentiometer 164 serve for manually con-

trodes 134, for the purpose of manually tuning the condenser 135 to receive any desired station. The arrangement shows the control electrodes quite independent of the condenser electrodes so that there is no coupling between the control circuit and the oscillator circuit. More broadly speaking the arrangement avoids coupling between the controlling and controlled circuits.

In one particular form of construction of the 10 condenser, the dielectric may comprise a piezoelectric crystal having the two condenser plates arranged on opposite surfaces thereof. By applying a potential, which may be applied to electrodes separate from the condenser plates, the 15 crystal may be deformed to vary the capacity between the two plates.

The methods of tuning by altering the inductance or capacity of an oscillatory circuit by varying an applied bias, lends itself particularly 20 to the remote control of radio receivers by means of a series of push-buttons which are so connected to points on the controlling resistance or potentiometer that a desired station may be received by merely depressing the corresponding push-button. Such push-button receivers were first described by me in my U. S. Patent No. 1,969,209, patented 7th August 1934.

One such embodiment is shown in Fig. 18 in which the remote tuning or other tuning of a receiver E is effected by varying at the control panel a regulating resistance R which alters the pre-magnetising current of one or more magnetic variometers serving for the tuning of the circuits, then a press-button control may be achieved by providing, in addition to the slider S on the regulating resistance R for effecting continuous tuning, further adjustable sliders having connections leading to the individual press-buttons  $D_i$ — $D_6$ .

By pressing in one of the press-buttons, the resistance corresponding to the setting of the slider will then be switched in, and the receiver tuned in to the station corresponding to the adjusted current or voltage. The advantage of this arrangement over ordinary press-button controls or press-button remote controls, resides in the fact that only one single conductor for the regulating current is required for an unlimited number of press-buttons and therefore of stations to which the set can be tuned. Furthermore, the possibility is provided of having a continuous change of wavelength available at the control panel in addition to a number of pre-selected stations. If there is provided a large number of permanently pre-selected stations, it is particularly expedient to provide stabilisers, voltage governors or the like for the sources of the regulating current.

If as tuning means there be employed devices which are substantially free of after-effects, no further complications of the arrangement are to be expected. This applies for example when valves are used as tuning elements, provided these valves serve as variable inductive or capacitative resistances. If tuning is effected with the aid of magnetic variometers, that is to say for instance by utilising saturation phenomena, or if mechanical driving means be employed, that is to say for instance driving means working on the principle of a voltmeter or ampere-meter, these means being adapted for instance to turn a rotary condenser to a greater or less extent in dependence on the adjustment of the regulating resistance. precautions must be taken to counteract these trolling the potential applied to the control elec- 75 after-effects, such as coercive force or friction.

The outcome of these after-effects may be eliminated with the aid of any means by which an increase of the relative accuracy is achieved. For this purpose there is contemplated more particularly the sub-dividing of the total frequency band to be covered by the set into a large number of sections, as described, for example, in British patent specification No. 409,737.

Assuming, for example that a remote-tuned set of this kind, designed to be tuned with a pre- 10 magnetisable self-inductance equipped with a ferro-magnetic core, that the normal broadcasting range of say 500 to 1500 kc/sec is to be covered, and assuming further that the registering error is reduced by the use of heavily alloyed transformer sheets to about 2 to 3%, then this means that at a medium frequency of the set range, that is to say, at about 1000 kc/sec, there will still be a difference of 20 to 30 kc/sec between the wave length to which a remote-controlled ar- 20 rangement adjusted by means of a regulating device is supposed to be adjusted and the wave length to which the set is actually adjusted. This, however, means a displacement through two to three receiving channels, with the result that 25an arrangement of this kind is bound to be equipped with a separate compensating indicating device or the like. By the use of nickle-ironalloys, such as "Permalloy" or the like, this registering error may, it is true, be reduced very con- 30 siderably further still, but the high price of materials of this kind sets a strict limit to its use.

If, however, the wave length range up to 1000 kc/sec be sub-divided, for instance, in such a manner as to provide five sub-ranges of about 35 200 kc each, the same registering error of about 2 to 3% in each sub-range will only amount to 4 to 6 kc/sec. This means that the minimum deviation from the mean value amounts to only 2 to 3 kc/sec. Now since a transmission chan- 40 nel has a breadth of 9 to 10 kc/sec, it follows that in spite of the registering error tuning will always be effected to within the required channel, and it is possible to calibrate the set correctly for particular stations inscribed by 45 name on the control panel without the use of additional compensating arrangements or the like, the control panel being provided for instance with a variable resistance or the like for the purpose of selecting the individual stations 50 within the sub-ranges.

Since a variable resistance of this kind allows of continuous, i. e. non graduated, tuning, any loss of quality in the reproduction due to the residual detuning of the order of 2 to 3 kc/sec 55 may be compensated for by merely turning the tuning knob at the control panel without, however, causing the index pointer coupled with the knob to move away from the position on the dial representing the station concerned.

If a set of this kind is provided with automatic sharp tuning, this latter will effect correct adjustment of the set without additional tuning, since the error lies within the limits of  $\pm 2 \pm 3$  a deviation with which automatic sharp 63 tuning arrangements can cope.

Arrangements which do not work perfectly continuously are capable of giving satisfactory results when using this principle. If, for example, the rotary condenser spindle of a receiver 70be driven by a ratchet wheel mechanism which in its turn is set in motion by means of a dialling disc or the like as in automatic telephone sets, it will be clear that with an arrangement of

wheel mechanism be made fine enough, or provided a sufficiently high transmission ratio is provided between the ratchet wheel mechanism and the condenser spindle, it would be possible to select each station with sufficient accuracy.

Such fine sub-division of the ratchet wheel drive would, however, necessitate an extremely large number of contacts for use in selecting a station. In this direction again a remedy is provided by the sub-dividing of the wavelength

If the total wavelength range be first subdivided into a number of sub-ranges, it becomes possible to select any desired station within these sub-ranges without an excessively large number of impulses, since the registering error due to the fact that the selecting of a station is no longer effected continuously but in small stages, becomes so small, as expressed in kc/sec, owing to the preceding sub-division of the wave length range into a considerable number of subranges, that the residual deviation can be compensated for by means of automatic tuning control.

It thus becomes possible in this case, in contrast to arrangements such as those used in automatic telephony where only a certain definite receiving point can be selected at a time owing to the use of preselectors and main selectors, by the use of automatic tuning control. to select any stations comprised within the subranges, the wavelengths of which need not be known at all at the time the set is built.

If in the first mentioned constructional example, the variable resistance be tapped at certain intervals and the tappings be connected to press-button contacts, or in the case of the last mentioned example if arrangements be provided which transmit a certain previously adjustable number of impulses, either magnetically or electromagnetically or otherwise when a pressbutton is operated, then the described arrangements may also be used in connection with pressbutton control. For this purpose the arrangement may be such that when a certain pressbutton is operated not only is a certain station selected within a sub-range, but the sub-range itself is also selected.

The sub-ranges may in their turn either be switched on likewise by means of ratchet wheel mechanisms, as indicated by way of example in Fig. 19 of the accompanying drawings, or they may be selected by means of separate leads connected severally to principle relays; or alternatively there may be provided arrangements working on a principle similar to that of the transmission of signalied orders on ships, whereby an indicating and a switching lever are always in corresponding positions in the receiver and 60 control instruments respectively. .

The described individual embodiments may also be combined in any desired manner.

Motor drives may also be employed for selecting the sub-range or for fine tuning within the sub-range, or for both purposes: in every case the advantage of sub-dividing the wavelength range which is achieved that any inaccuracies resulting from the running on of the motor due to inertia or to idle movement in the gearing used, and any registering errors due to these causes are rendered harmless in their effects by the sub-division of the wavelength range, it being possible to use automatic tuning devices for compensating the residual errors. this kind, provided the teeth of the ratchet 75 Since with the said arrangements it is essential

323,450 **15** 

to avoid registering errors, it is desirable, in order to obtain the desired result with a minimum number of sub-ranges, to take steps to ensure that the relative inaccuracy within the individual sub-ranges remains the same. In an arrangement providing for graduated switching of the sub-ranges of the type shown in Fig. 19, care must therefore be taken to ensure that the sub-ranges have as nearly as possible an approximately equal extent expressed in kc/scc. 10

If in the case of a set in which fine tuning is effected by saturation, and in which therefore the inductivity is variable gradually and not in steps (which may be effected not only by means of premagnetising but equally well, as in the case of varying capacity, by the use of capacitive or inductive apparent resistances with the aid of appropriate valve circuits), the switching of the ranges being effected solely by connecting larger or smaller condensers to this variable inductance, 20 then this precondition will be only very imperfectly fulfilled. This is the case since when larger or smaller condensers are connected in, the range of variation of the inductivity, expressed as a percentage, remains always the same, from 25 which it follows that when small condensers are connected in, on alteration of the inductivity in certain proportion, very much greater wavelength ranges are swept through than when larger condensers are connected in, owing to the 30 lower wave ranges to which adjustment is effected when the smaller condensers are connected in.

To overcome this difficulty the circuit arrangement shown in Fig. 19 is provided, which shows only a single oscillating circuit suitable for use 35 for example in a single span superheterodyne set. If a superheterodyne receiver with one or more preselection circuits be employed the tuning elements shown in Fig. 19 must be provided in an appropriately larger number.

This applies to a still greater extent to multiclrcuit in-line receivers, which however are for other reasons less suitable for the purposes of the present invention although their use is not intended to be precluded.

LI in Fig. 19 denotes the variable self-inductance proper. In the position shown in the drawing, LI is connected to the condenser CI to form an oscillating circuit and in this case the entire range of operation of the coil Li is utilised. 50 When the contact segments are advanced to the extent of one contact by means of the ratchet wheel mechanism St, the coil Li is coupled with the condenser C2 to form an oscillating circuit. Owing to the relatively small sub-ranges (which 55 incidentally affords the further advantage that the range of variation of the variometer LI used or of corresponding capacity varying means can be kept small) the amount of transmission channels covered in the second range will not be substantially different compared to the first range. although C2 should be somewhat smaller than CI so that the sub-range corresponding to the next frequencies is then switched on.

If however on further movement of the switching arm to the third contact, the next higher range be then switched in, the range of frequency would nevertheless be gradually increased to such an extent that the pre-condition set forth above as being desirable can no longer be regarded as then being approximately fulfilled. To overcome this drawback there is employed the switching arm B. This switching arm connects in parallel with the coll Li a coil L2 which has the effect of causing the alteration of inductivity 75

to be less in the next two switching positions than in the preceding switching positions, since the total inductivity of the circuit then consists of a fixed inductivity L2 with which a variable inductivity L1 is connected in parallel.

On further reducing the frequency by switching over to fresh ranges there are then switched in a first tapping and later on possibly one or more subsequent tappings of the coil L2. Although, owing to the switching operations affecting the coil L2 a reduction of wavelength likewise occurs, whereby the switching over of condensers might be avoided or the number of condensers to be switched in by means of the switching arm A might be reduced, it is nevertheless advisable in many cases to provide separate condensers for each individual contact of the switching arm A, since it is possible by adjustment of the condenser CI, C2 and Cn to effect a synchronising equilisation between the individual circuits of the set in the case of muticircuit sets. Connecting a fixed coil L2 in parallel with the variable inductance Li has the further advantage that the quality of the coil may thereby be adapted to the individual wavelengths.

If, for example, coils having ferro-magnetic coils be used, it is possible, for the long wave ranges, to employ coils having cores of relatively high permeability, whereby inter alia the expenditure involved in pre-magnetising can be reduced; when the wavelengths are decreased, however, coils of this kind would exert a very considerable damping action. Now since the coil L2 is connected in parallel with them, and this coil becomes more and more decisive for the damping of the circuit as the length of the received waves becomes progressively shorter, this effect can be kept within favourable limits.

What is here said in regard to variation of inductivity also applies, with appropriate modification, to arrangements working with alteration of capacity for tuning purposes, as also to all the other embodiments mentioned above, as in fact all the individual embodiments discussed in the present application may be combined in any desired manner, it being understood that all such combinations fall within the scope of the invention for which protection is sought.

The connecting of the controls of the sct to the set proper may be effected in a variety of ways. Either separate leads may be provided or existing conductors, for instance, the heavy current conductor may be used for control purposes by superposing high frequency currents or low frequency currents on the conductor current, or alternatively by tapping direct currents in alternating current leads with the aid of the point switching arrangements. The embodiment of a direct high frequency radiation known per se between the control device and the receiving or transmitting set to be controlled, for the purpose of directing the controlling and regulating currents, is likewise possible. In this connection, for the purpose of varying amplitude or frequency, continuous regulation of a capacitive or inductive tuning element associated with the receiving or transmitting set to be controlled may be effected, while the switching over from one wave length range to another, and the like, may be brought about by varying the number of transmitted impulses.

overcome this drawback there is employed the switching arm B. This switching arm connects in parallel with the coll L1 a coil L2 which has the effect of causing the alteration of inductivity 75

In these cases, however, it is necessary to provide a lead for transmitting the sound output from the receiving set to a control device when the receiving set is not in the same place or room

as the controlling device, but if the remote control is to afford all its advantages, it is expedient to set up the receiving device remote from where the control device is operated, for instance, from a dwelling room but even in a case of this kind it is possible to do without a connecting conductor provided, for instance, by means of a short wave or ultra short wave connection maintained in a certain fixedly adjusted canal, the transmission received by the receiving set can be radiated to the place from which it is desired to hear it, where it can then be received by a receiver which is adjusted to a fixed wavelength and therefore does not require any additional controlling means.

In this case also, the radiation may be effected along existing leads, for instance, along mains current or power current leads. Installations of this kind are particularly suitable when it is desired to receive various transmissions in 20 several rooms, and it is then possible for various reproducing sets, to which the programmes are fed in different short wave channels or even long wave channels, to be operated in conjunction with different receivers proper remote controlled 25 with or without wiring connections and placed for instance in a central or storage room. The short wave retransmitting of a received transmission to the reproducing device proper affords particularly great advantages in the case of sets 30 to which only a limited number of persons are required to listen-in. After a transmission received by a receiving set and particularly by a remote control receiving set be retransmitted on an ultra-short wave which is intended to be 35 divisions [--10 of the scale. fixedly or at least practically fixedly adjusted, it becomes possible to provide the individual occupants of a house with headphones on which a small ultra-short wave receiver is directly mounted or otherwise combined in a simple form. In 40 the case of a valve receiver, it is generally sufficient, since the transmitting distance is but slight, to provide a single valve which may be operated by means of a pocket lamp battery.

In many cases, however, a detector set will 45 suffice, say that for example when a one or two meter wave is used merely a short length of wire on the headphone will suffice as the receiving antenna which feeds its energy to a crystal detector or the like which converts the ultra-short 50 wave energy into sound that can be directly heard in the headphones.

Sound generators are also known which convert high frequency outputs directly in the sound output with the aid of a hearing wire which is 53 extremely thin and therefore has a negligible amount of thermal inertia. If the wire of a thermal telephone of this kind, which as is wellknown can be made extremely small and light so that it can be introduced into the ear passage in the form of a small plug, be directly inserted in the antenna circuit of an ultra-short wave receiver, the complete arrangement will consist of a conductor which will be of a length of about half a meter when the wavelength used is one 65 meter, the wire of the thermal microphone being connected approximately to the middle of this conductor.

This arrangement or a similar arrangement enables the receiving apparatus to be extremely 70 light and small, even smaller than an ordinary headphone, and enables the wearer of such a headphone to receive broadcast programmes in any of the rooms of a house equipped with the described installation without other persons in 75

these rooms being in any way disturbed by these programmes.

The employment of this principle is not restricted to the arrangements described in the present application, although it becomes particularly advantageous in connection therewith.

The control device for one of the arrangements described above may take the form illustrated in Fig. 20. On the left is shown a dialling disc serving for the selection of any one of for instance ten sub-ranges. When a subrange has been dialied the appropriate wave of the divisions I to 10 on the scale Sk is illuminated, which may be effected by means of contact mechanism coupled with the dialling disc, which causes the division of the scale concerned to remain illuminated until the set is switched off or a new sub-range selected. For this purpose there is coupled with the dialling disc W a contacting mechanism which, on the dialling of a number, carries a contact arm or the like along with it and leaves this arm in its switching position until a new number is dialled, whereupon it is restored to its position of rest. as the mechanism of the dialling operation, by means of spring force or other agency and then on the dialling of a new number assumes a fresh position corresponding to the new number dialled.

As illuminating means there may suitably be used special glow lamps which may, for instance, be constructed after the manner of the known amplitude valves and which may be placed in suitable positions beenath each of the

Ki denotes a knob coupled to a potentiometer or variable resistance whereby the fine adjustment of the stations within a sub-range is effected. Coupled with the knob KI by means of a cord or the like is a pointer Z which can be moved up and down as indicated by the double arrow, thereby indicating which of the stations marked up in the divisions i-10 has been selected.

K2 denotes a loudspeaker regulating potentiometer or variable resistance which may be inserted in the lead to the loudspeaker, or which may be arranged to vary the grid bias of one or more valves in the receiver for the purpose of volume control, or which may effect the volume control in any other known manner.

There is further provided at any suitable point in the control device a switch which controls, either directly or through the intermediary of a relay, the switching on and off of the current supply which will generally be taken from the mains.

Instead of the dialling disc W it is also possible to provide beneath the divisions !— 10. 60 press buttons which when operated select a sub-range through the intermediary of appropriate relays, or alternatively, when operated cause, with the aid of a mechanical or electromagnetic mechanism, a number of impulses to be transmitted which in their turn effect the selecting of a particular sub-range in the receiving set.

As selectors for the switching of the individual wavelength ranges there may be employed conventional pre-selectors such as are commonly used in automatic telephone installations. The same applies also to the dialling discs used in the control device.